

HBT's thermal impedance measurement

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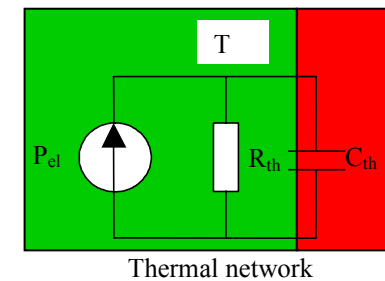
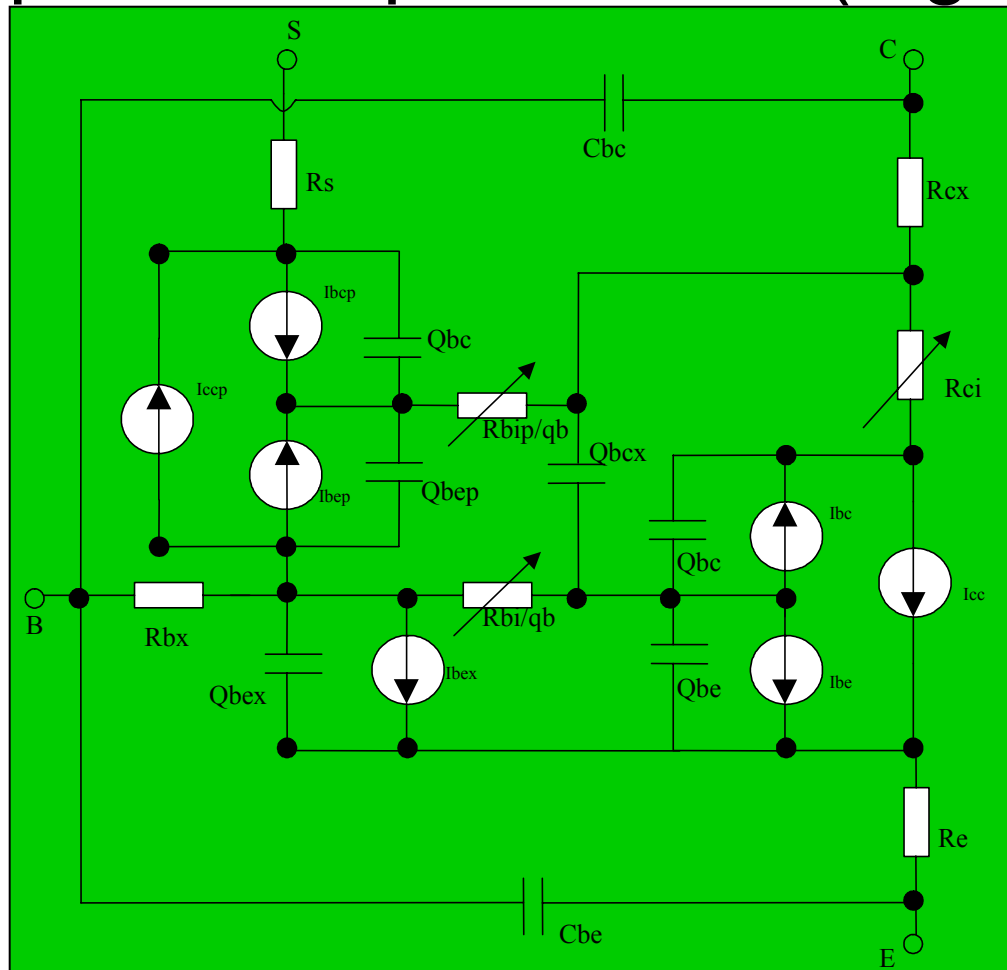
♣ University Bordeaux 1, F

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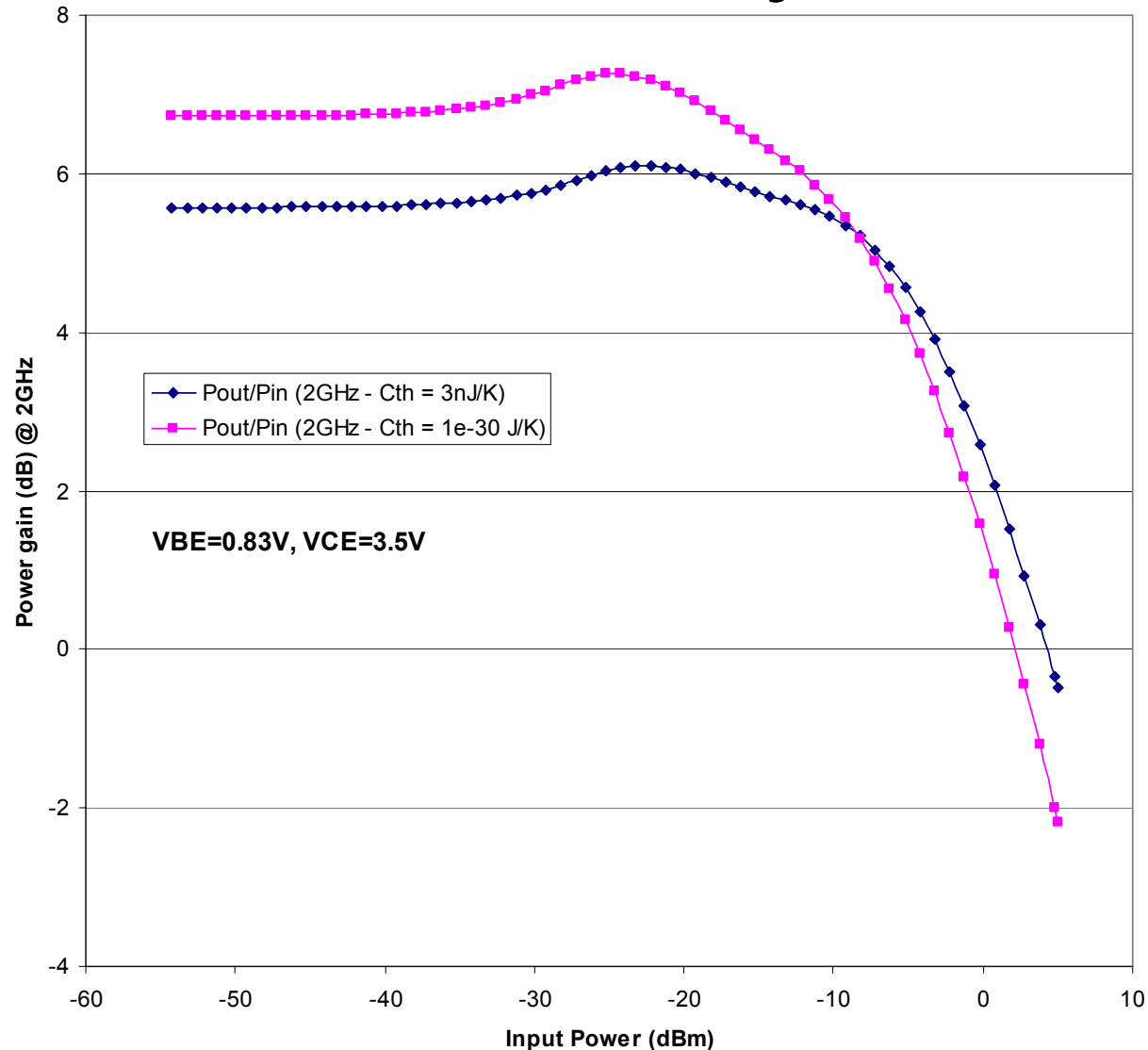
♦ ST Microelectronics, Crolles, F

Motivation

- Bipolar compact model (e.g. VBIC)



Do we really need C_{TH} ?



Harmonic balance
simulation:

HICUM with ADS

RF power transistor

Power gain as a
function of input
power at 2GHz

We do really need C_{TH} .

- But, how to get it ?

Outline

- Motivation: done
- Presentation of two methods for:
 - Z_{TH} measurement
 - C_{TH} extraction
 - Z_{TH} modelling
- Scaling issues
- Questions

Z_{TH} measurement

1. Load Pull measurement
2. Small Signal measurement
 - frequency range
 - NWA
 - Z-parameters
3. Pulsed measurement

Load Pull measurement

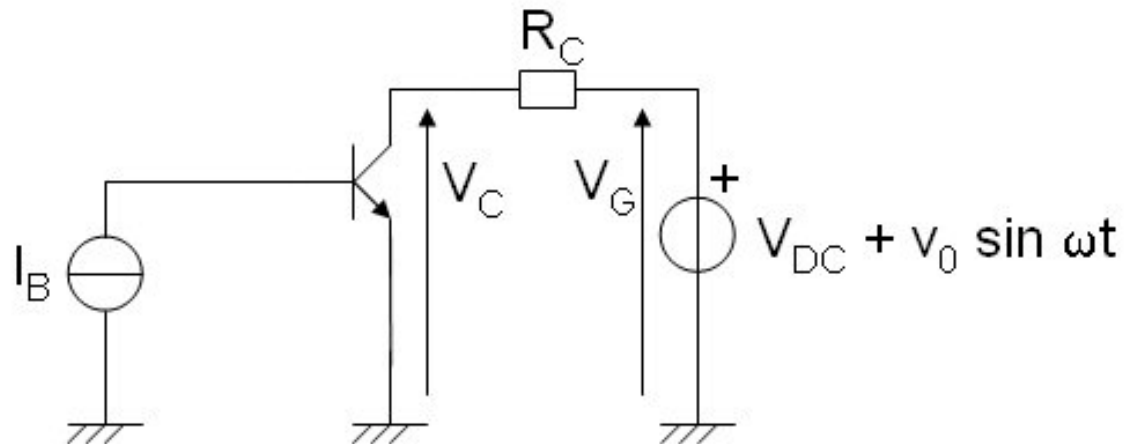
- Measurement equipment
- De-embedding and calibration
- Indirect extraction
 - First, full compact model parameter extraction
 - Second, C_{TH} fitting
 - Third, C_{TH} is in general not very sensitive
- => For instance, not the best choice

Small Signal measurement

- Frequency range: first order guess
 - Standard transistor:
 - Deep trench isolation
 - $0.25\mu\text{m} \times 12.8\mu\text{m}$
 - $R_{\text{TH}} = 1000 \text{ K/W}$
 - $C_{\text{TH}} = 0.5 \text{ nJ/K}$
 - $f_{\text{TH}} = 1/(2\pi R_{\text{TH}} C_{\text{TH}}) \approx 300 \text{ kHz}$
 - Out of the range of standard NWA

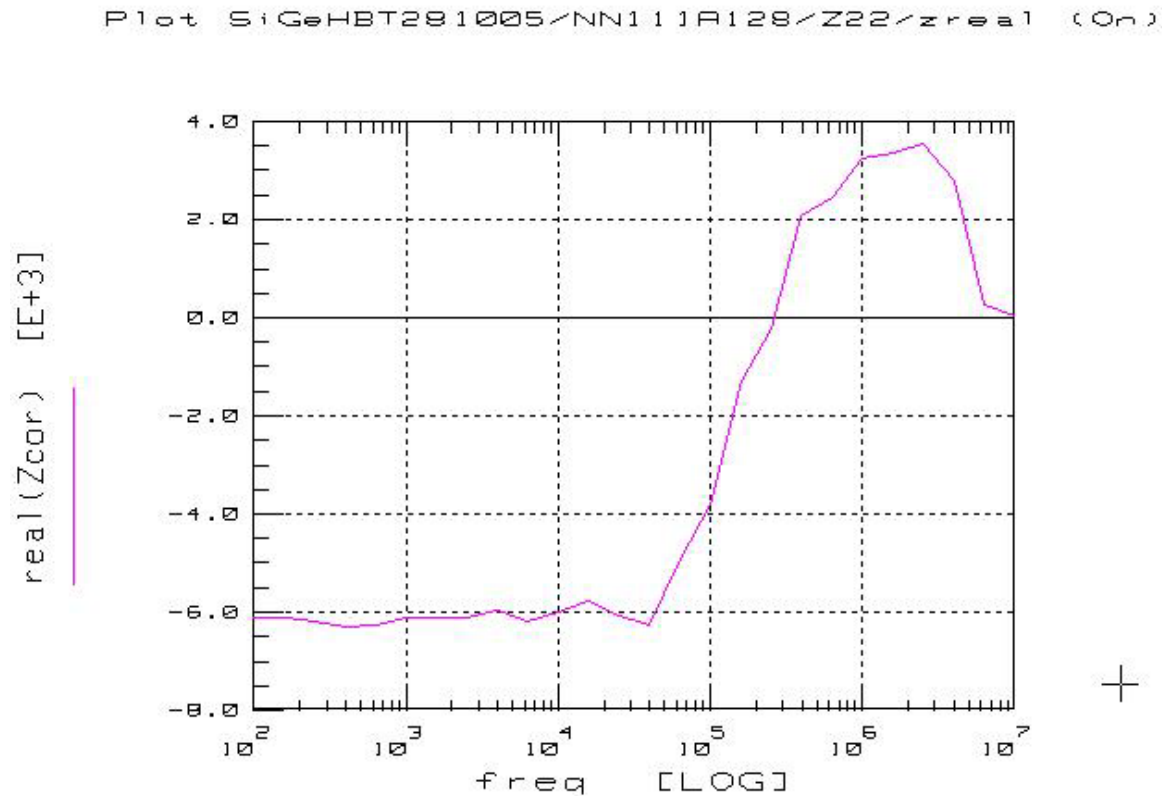
Measurement setup: AC

- Base: I_B constant
- Collector: $V_{DC} + V_{AC}$
- I_C through R_C
- $Z_{CE} = V_C / I_C$ (complex)
- frequency range
 - Theory: $0 \rightarrow \infty$
 - $0.1\text{Hz} \rightarrow 10\text{ MHz}$



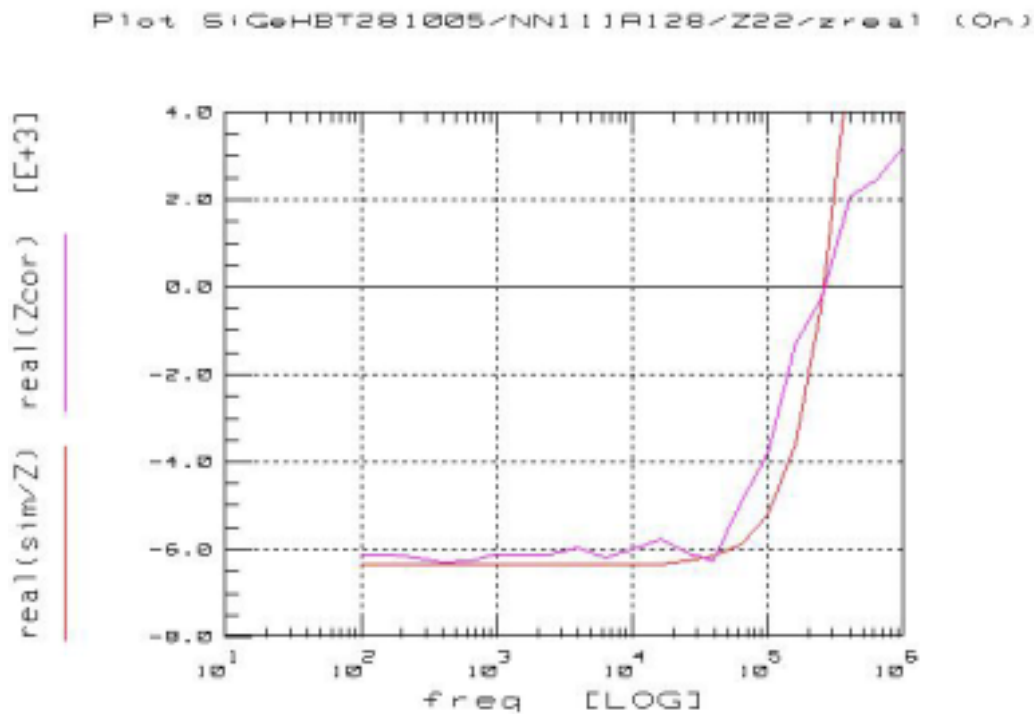
Measurement results: AC

- Real Part of Z_{CE}



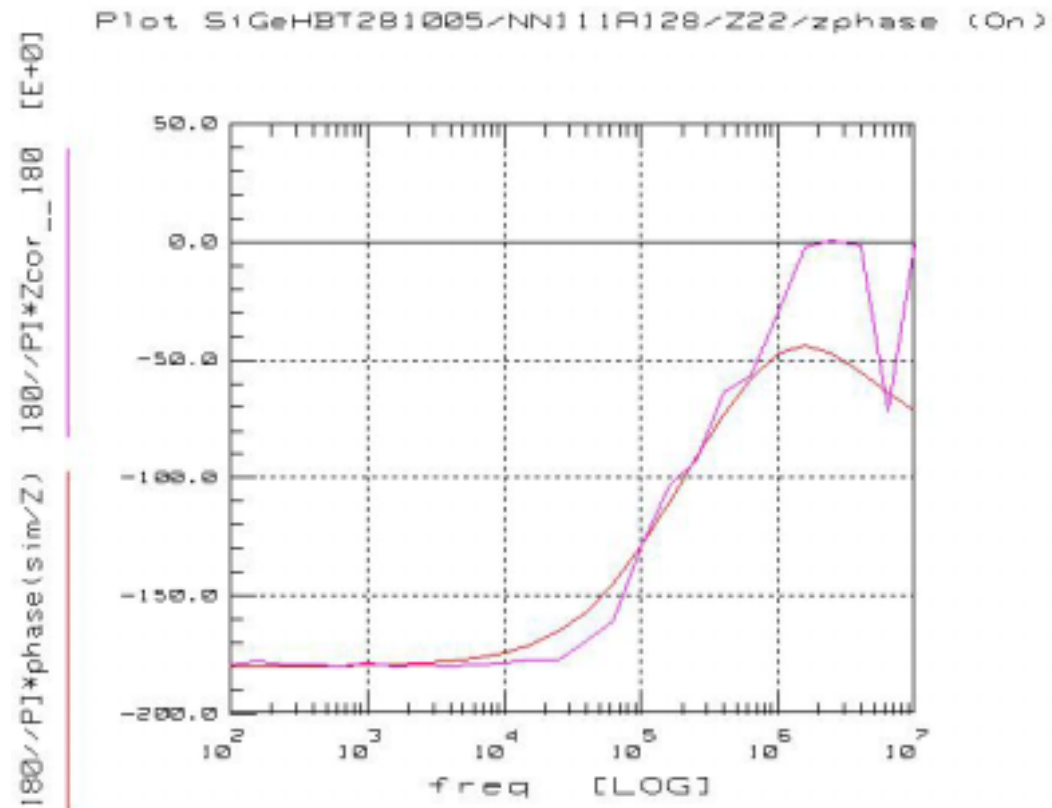
Modelling: AC

- Hicum compact model: R_{TH}
 - Real part of Z_{CE} at low frequencies



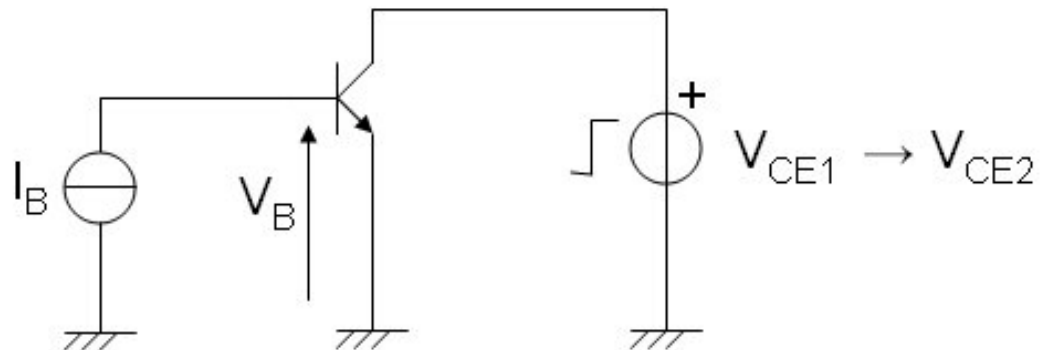
Modelling: AC

- Hicum compact model: C_{TH}
 - Phase of Z_{CE}



Measurement setup: pulse

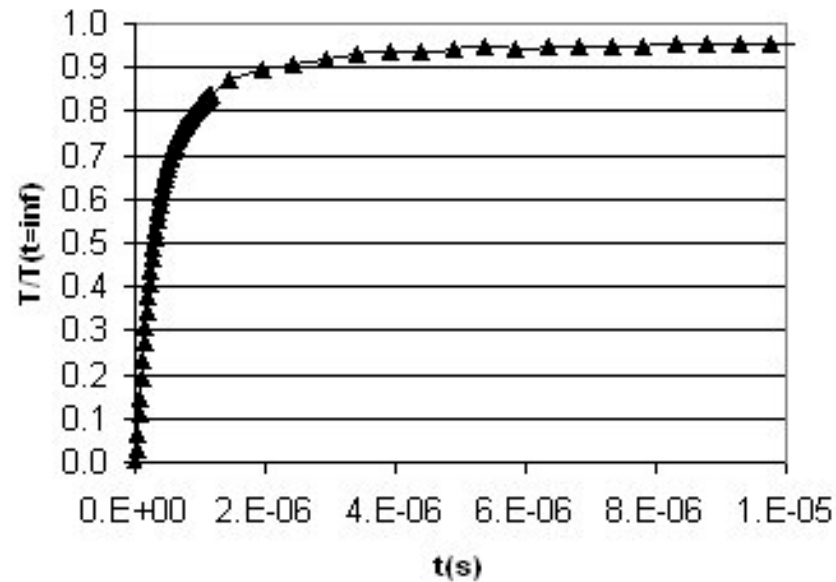
- Base: I_B constant
- Collector: $V_{CE1} \rightarrow V_{CE2}$
- Measure V_B
- Calibration $V_B \rightarrow T$
- time range
 - Theory: $0 \rightarrow \infty$
 - $10\text{ns} \rightarrow 10\text{s}$



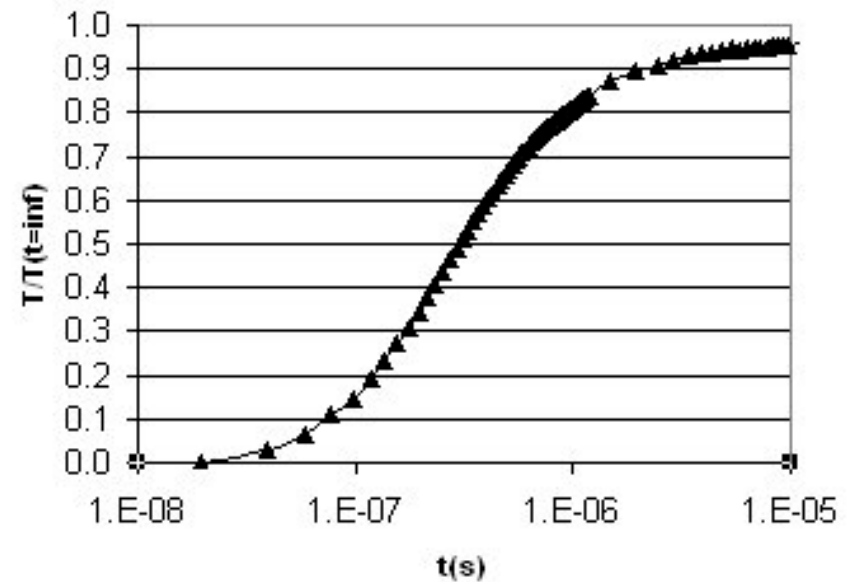
Measurement results: pulse

- $V_{BE}(t) \rightarrow T(t)$

Lin

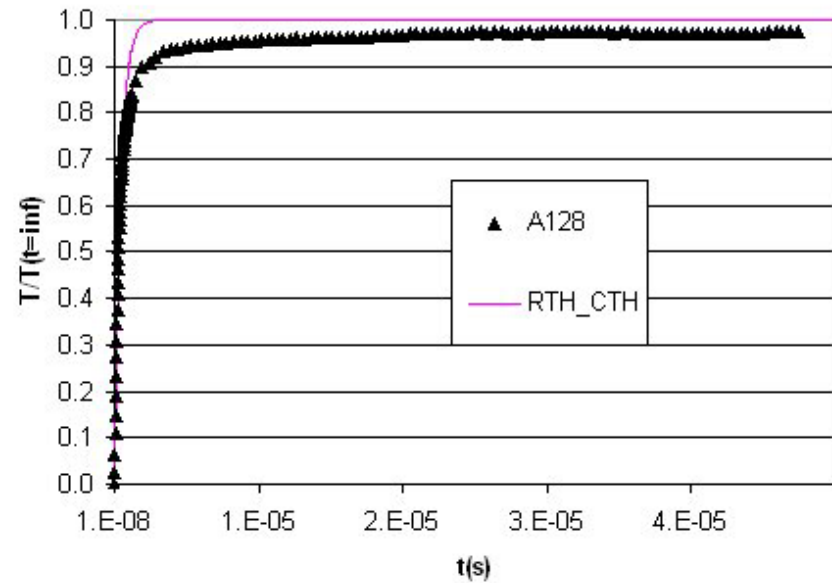


Log

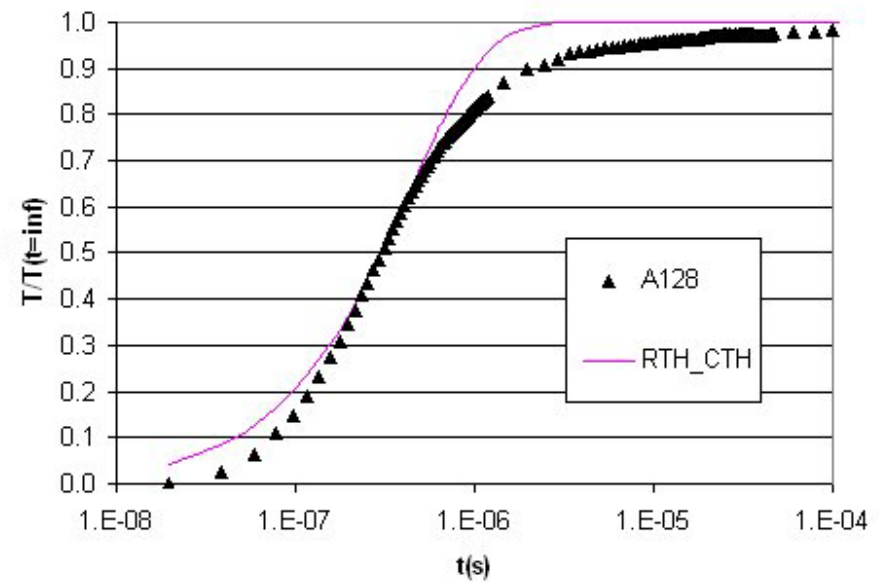


Modelling: pulse

- R_{TH} - C_{TH} cell:
lin

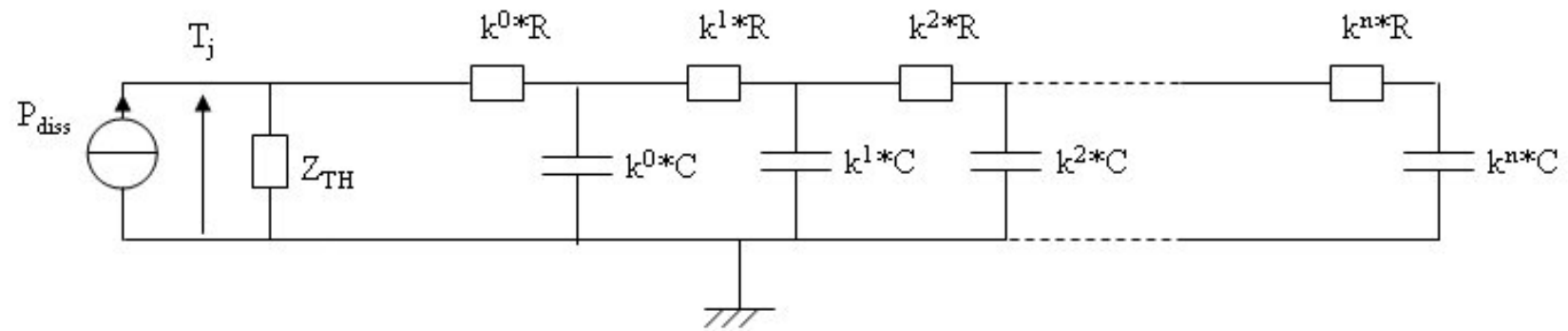


log



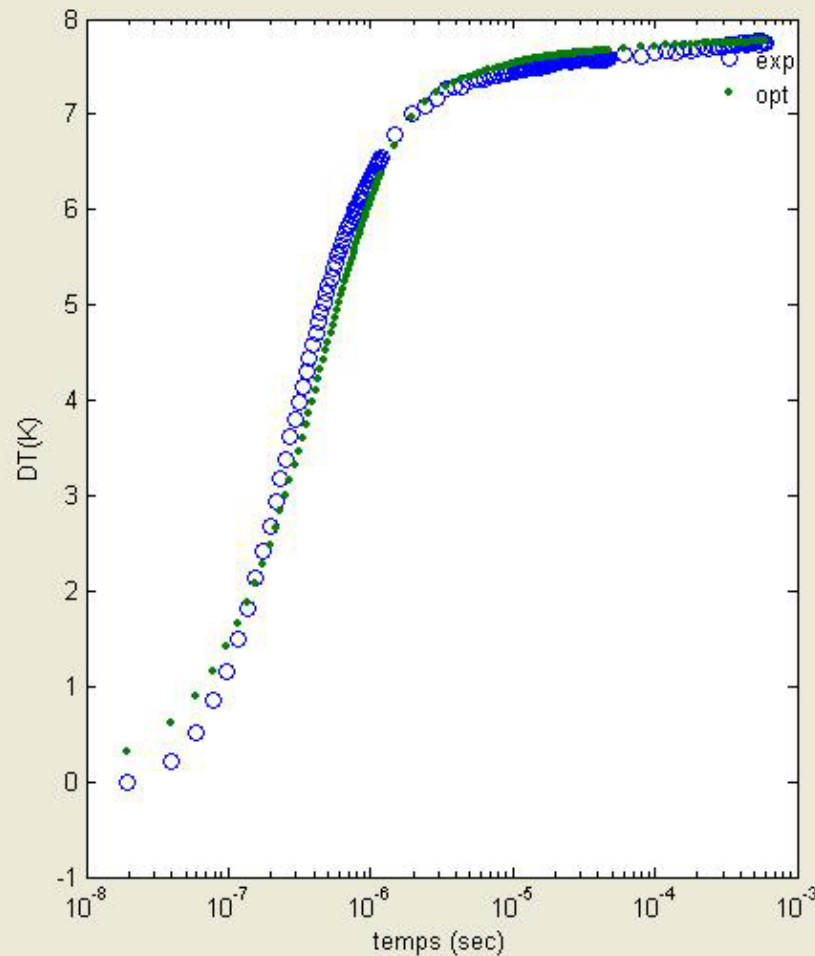
Modelling pulse: recursive network

- Recursive network



Recursive network: toolkit & results

Resultat



Flux (mW) L1 (μm) W1 (μm)

New file load:

Plotting choice:

Recursive ($n \geq 1$ and $k > 0$)
☒ Yes ☐ No

Thermal Recursive Impedance

Identification

☐ "O" Position from HBT_SiGe

☐ Asymptotic values

☒ RTH (K/W)

☐ CTH (J/K)

☐ Thermal conductivity

☐ "Backend" (W/m/k)

☐ Ct (J/K)

Identification Parameters

k : (recursive number)

n: cell number

Time Range

Initial time (s)

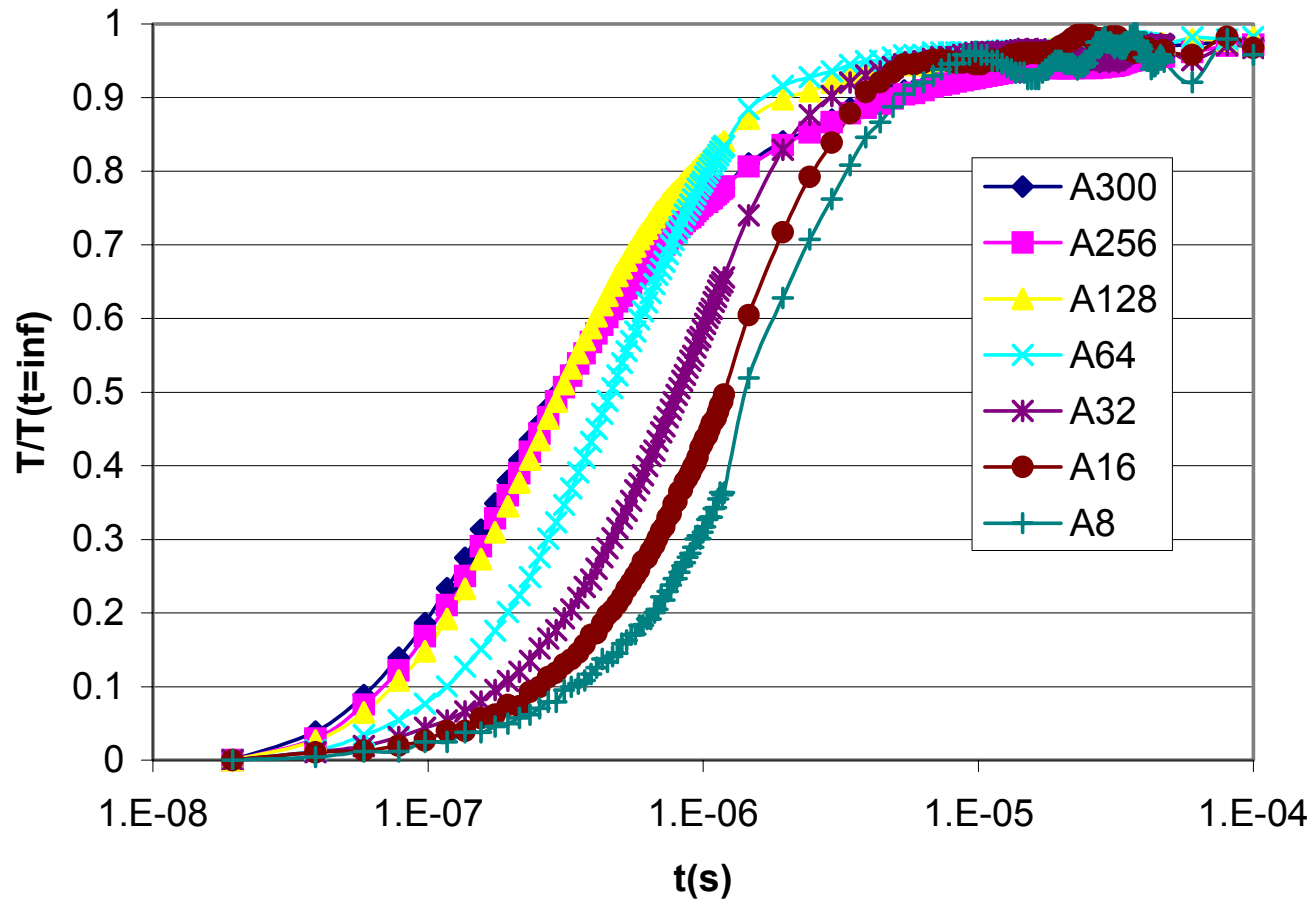
Last time (s)

Store results >>

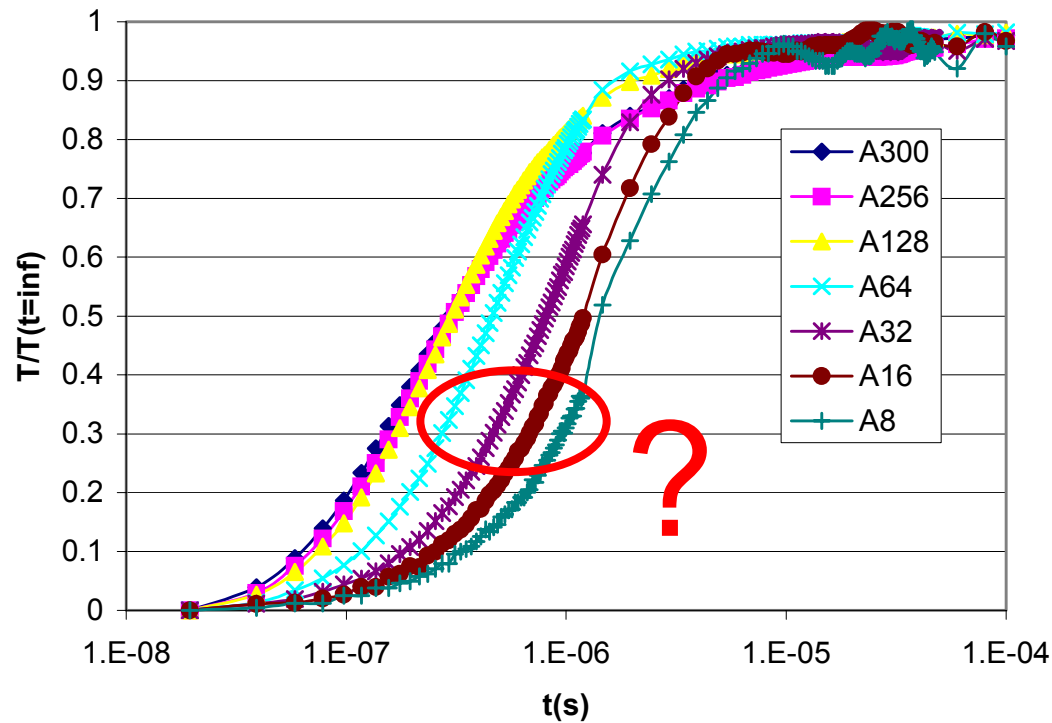
Listbox

Name	RTH (K/W)	k	CTH (J/K)
RC	1393.119	0	3.3905e

Scaling issues

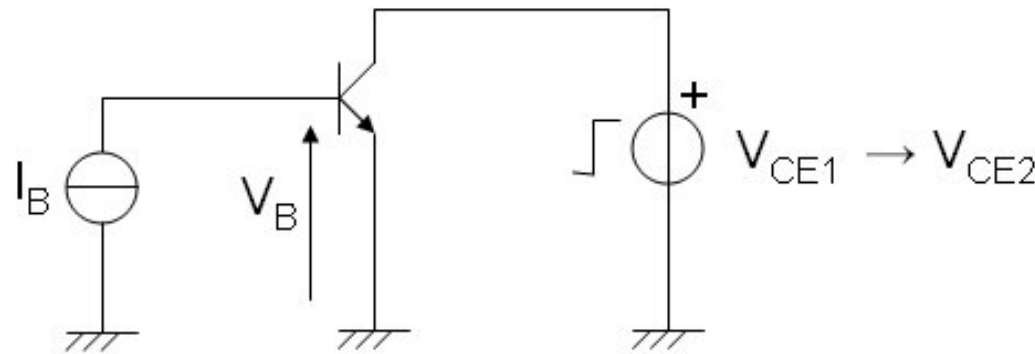


Why does the dynamic behaviour change for small devices?



Measurement problem

- Small transistors, small I_B , high Z_{in}
- New current source: new dynamic behaviour !!!



- Solution : ?

Discussion

- Pulse and AC method for C_{TH} determination
 - Both methods give similar results
 - Both methods need similar equipment
 - AC: a full parameter extraction has to be performed for R_{TH} and C_{TH} determination
 - Pulse: R_{TH} and C_{TH} can be directly determined, but a V_{BE} -T calibration is necessary
- Recursive network for accurate modelling
- Limits for small transistors

- Thanks for your attention